

100V, High-Side, Current-Sense Amplifiers with Voltage **Output**

1 FEATURES

- Ideal for High-Voltage Current Monitoring **Applications**
	- Wide 5V to 100V Input Common-Mode Range
	- Independent Operating Supply Voltage
- ⚫ High Accuracy and Low Quiescent Current Support Precision Application Requirements
	- ±0.1% Full-Scale Accuracy
	- Low 100μV Input Offset Voltage
	- Four Gain Versions Available
		- 20V/V (RSA4080A/RSA4081A)
		- 50V/V (RSA4080B/RSA4081B)
		- 60V/V (RSA4080C/RSA4081C)
		- 100V/V (RSA4080D/RSA4081D)
	- 106μA Supply Current (RSA4080)
	- 115μA Supply Current (RSA4081)
- ⚫ Flexible Current Sensing Supports Monitoring of Charge and Discharge of Batteries
	- Bidirectional (RSA4081) or Unidirectional (RSA4080) ISENSE
	- Reference Input for Bidirectional OUT (RSA4081)
- ⚫ Micro Size Packages: SOP8, MSOP8

2 APPLICATIONS

- Automotive (12V, 24V, or 42V Batteries)
- ⚫ 48V Telecom and Backplane Current **Measurement**
- ⚫ Bidirectional Motor Control
- ⚫ Power-Management Systems
- ⚫ Avalanche Photodiode and PIN-Diode Current Monitoring
- ⚫ General System/Board-Level Current Sensing
- ⚫ Precision High-Voltage Current Sources

3 DESCRIPTIONS

The RSA408X are high-side, current-sense amplifiers with an input voltage range that extends from 5V to 100V making them ideal for telecom, automotive, backplane, and other systems where high-voltage current monitoring is critical. The RSA4080 is designed for unidirectional current-sense applications and the RSA4081 allows bidirectional current sensing. The RSA4081 single output pin continuously monitors the transition from charge to discharge and avoids the need for a separate polarity output. The RSA4081 requires an external reference to set the zero-current output level (V_{SENSE} = 0V). The charging current is represented by an output voltage from V_{REF} to V_{CC}, while discharge current is given from V_{REF} to GND.

For maximum versatility, the 100V input voltage range applies independently to both supply voltage (V_{CC}) and common-mode input voltage (V_{RS+}). Highside current monitoring does not interfere with the ground path of the load being measured, making the RSA408X particularly useful in a wide range of highvoltage systems.

The combination of four gain versions (20V/V, 50V/V, 60V/V, $100V/V=$ A, B, C, D suffix) and a userselectable, external sense resistor sets the full-scale current reading and its proportional output voltage. The RSA408X offer a high level of integration, resulting in a simple, accurate, and compact currentsense solution.

The RSA408X is available in Green SOP8 and MSOP8 packages. It operates over an ambient temperature range of -40°C to 125°C.

Device Information (1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
RSA408X	SOP8	4.90mm x 3.90mm
	MSOP8	3.00mm×3.00mm

For all available packages, see the orderable addendum at the end of the data sheet.

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4 REVISION HISTORY

Note: Page numbers for previous revisions may different from page numbers in the current version.

5 PACKAGE/ORDERING INFORMATION (1)

NOTE:

(1) This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the right-hand navigation.

(2) There may be additional marking, which relates to the lot trace code information (data code and vendor code), the logo or the environmental category on the device.

(3) RUNIC classify the MSL level with using the common preconditioning setting in our assembly factory conforming to the JEDEC industrial standard J-STD-20F, Please align with RUNIC if your end application is quite critical to the preconditioning setting or if you have special requirement.

6 PIN CONFIGURATION AND FUNCTIONS

Pin Description

 (1) I = Input, O = Output, P=Power.

7 SPECIFICATIONS

7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)⁽¹⁾

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

(2) Short-circuit to ground, one amplifier per package.

(3) The package thermal impedance is calculated in accordance with JESD-51.

(4) The maximum power dissipation is a function of TJ(MAX), RθJA, and T_A. The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A) / R_{\theta JA}$. All numbers apply for packages soldered directly onto a PCB.

7.2 ESD Ratings

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

ESD SENSITIVITY CAUTION

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

7.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

7.4 Electrical Characteristics

(V $_{\rm CC}$ = V $_{\rm RS+}$ = 5V to 100V, V $_{\rm REF1A}$ = V $_{\rm NEF1B}$ = 5V, V $_{\rm SSES}$ = (V $_{\rm RS+}$ - V $_{\rm RS}$.) = 0V, unless otherwise noted. Typical values are at T $_{\rm A}$ = +25°C, Full = -40°C to 125°C.) (1) (2)

NOTE:

(1) All devices are 100% production tested at $T_A = +25$ °C. All temperature limits are guaranteed by design.

(2) V_{REF} is defined as the average voltage of V_{REF1A} and V_{REF1B}. REF1B is usually connected to REF1A or GND. V_{SENSE} is defined as V_{RS+} - V_{RS-}.

(3) Limits are 100% production tested at 25°C. Limits over the operating temperature range are ensured through correlations using statistical quality control (SQC) method.

(4) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration.

(5) The common-mode range at the low end of 5V applies to the most positive potential at RS+ or RS-. Depending on the polarity of V $_{\text{SENSE}}$ and the device's gain, either RS+ or RS- can extend below 5V by the device's typical full-scale value of V_{SENSE}.

(6) Negative V_{SENSE} applies to RSA4081 only.

(7) VSENSE IS:

RSA4080A, 10mV to 250mV RSA4080B, 10mV to 120mV RSA4080C, 10mV to 100mV RSA4080D, 10mV to 50mV RSA4081A, -125mV to +125mV RSA4081B, -60mV to +60mV RSA4081C, -50mV to +50mV RSA4081D, -25mV to +25mV

(8) For RSA4080 V_{OS} is measured as (V_{OUT}/A_V) - V_{SENSE} at V_{SENSE}=10mV. For RSA4081 V_{OS} is measured as (V_{OUT} - V_{REF})/A_V at V_{SENSE} = 0V. (9) VSENSE is:

RSA4080A/RSA4080B/RSA4080C/RSA4080D, 10mV RSA4081A/RSA4081B/RSA4081C/RSA4081D, 0V $V_{REF1B} = V_{REF1A} = 2.5V$

- (10) For RSA4080, Output voltage is internally clamped not to exceed the lesser of +11.5V or Vcc For RSA4081, Output voltage is internally clamped not to exceed the lesser of +18V or V_{CC}
- (11) V_{REF} range should be between GND+1.5V and V_{CC} -1.5V, and MAX is 6V.
- (12) Output settles to within 1% of final value.
- (13) The device will not experience phase reversal when overdriven.

7.5 Typical Characteristics

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

Figure 6. Gain Accuracy vs Temperature

Typical Characteristics

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

Figure 9. V_{OUT} Low Voltage vs I_{OUT} (Sinking) Figure 10. Small-Signal Gain vs Frequency

Figure 11. Small-Signal Gain vs Frequency Figure 12. Small-Signal Transient Response

Typical Characteristics

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

Figure 13. Large-Signal Transient Response Figure 14. Small-Signal Transient Response

Figure 17. Large-Signal Transient Response

Figure 15. Large-Signal Transient Response Figure 16. Small-Signal Transient Response

Typical Characteristics

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

Figure 21. STARTUP DELAY

Figure 19. Positive Overload Recovery Figure 20. Negative Overload Recovery

8 DETAILED DESCRIPTION

The RSA408X unidirectional and bidirectional high-side, current-sense amplifiers feature a 5V to 100V input common-mode range that is independent of supply voltage. This feature allows the monitoring of current out of a battery as low as 5V and also enables high-side current sensing at voltages greater than the supply voltage (V_{cc}). The RSA408X monitors current through a current-sense resistor and amplifies the voltage across the resistor. The RSA4080 senses current unidirectionally, while the RSA4081 senses current bidirectionally.

The 100V input voltage range of the RSA408X applies independently to both supply voltage (V_{cc}) and commonmode, input-sense voltage (VRS+). High-side current monitoring does not interfere with the ground path of the load being measured, making the RSA408X particularly useful in a wide range of high-voltage systems.

Battery-powered systems require a precise bidirectional current-sense amplifier to accurately monitor the battery's charge and discharge. The RSA4081 charging current is represented by an output voltage from VREF to Vcc, while discharge current is given from VREF to GND. Measurements of OUT with respect to VREF yield a positive and negative voltage during charge and discharge, as illustrated in Figure 22 for the RSA4081A.

8.1 Current Monitoring

The RSA4080 operates as follows: current from the source flows through R_{SENSE} to the load (Figure 21), creating a sense voltage, V_{SENSE}. Since the internal-sense amplifier's inverting input has high impedance, negligible current flows through RG2 (neglecting the input bias current). Therefore, the sense amplifier's inverting input voltage equals Vsource - (ILOAD)(RSENSE). The amplifier's open-loop gain forces its noninverting input to the same voltage as the inverting input. Therefore, the drop across RG1 equals V_{SENSE}. The internal current mirror multiplies IRG1 by a current gain factor, β, to give $I_{A2} = \beta \times I_{RG1}$. Amplifier A2 is used to convert the output current to a voltage and then sent through amplifier A3. Total gain = 20V/V for RSA4080A, 50V/V for the RSA4080B, 60V/V for the RSA4080C, and 100V/V for the RSA4080D.

The RSA4081 input stage differs slightly from the RSA4080 (Figure 22). Its topology allows for monitoring of bidirectional currents through the sense resistor. When current flows from RS+ to RS-, the RSA4081 matches the voltage drop across the external sense resistor, R_{SENSE}, by increasing the current through the Q1 and RG1. In this way, the voltages at the input terminals of the internal amplifier A1 are kept constant and an accurate measurement of the sense voltage is achieved. In the following amplifier stages of the RSA4081, the output signal of amplifier A2 is level- shifted to the reference voltage (VREF = VREF1A = VREF1B), resulting in a voltage at the output pin (OUT) that swings above VREF voltage for positive-sense volt-ages and below VREF for negativesense voltages. Vout is equal to VREF when VSENSE is equal to zero. Set the full-scale output range by selecting RSENSE and the appropriate gain version of the RSA408X.

8.2 External References (RSA4081)

For the bidirectional RSA4081, the V_{OUT} reference level is controlled by REF1A and REF1B. V_{REF} is defined as the average voltage of VREF1A and VREF1B. Connect REF1A and REF1B to a low-noise, regulated voltage source to set the output reference level. In this mode, V_{OUT} equals V_{REF1A} when V_{SENSE} equals zero (see Figure 23). Alternatively, connect REF1B to ground, and REF1A to a low-noise, regulated voltage source. In this case, the output reference level (VREF) is equal to VREF1A divided by two. V_{OUT} equals VREF1A/2 when V_{SENSE} equals zero. In either mode, the output swings above the reference voltage for positive current-sensing ($V_{RS+} > V_{RS}$). The output swings below the reference voltage for negative current-sensing ($V_{RS+} < V_{RS-}$).

9 APPLICATIONS INFORMATION

9.1 Choosing the Sense Resistor

Choose RSENSE based on the following criteria:

Voltage Loss: A high RSENSE value causes the power-source voltage to degrade through IR loss. For minimal voltage loss, use the lowest R_{SENSE} value.

- Accuracy: A high RSENSE value allows lower currents to be measured more accurately. This is due to offsets becoming less significant when the sense voltage is larger. For best performance, select R_{SENSE} to provide approximately 250mV (gain of 20V/V), 120mV (gain of 50V/V), 100mV (gain of 60V/V) or 50mV (gain of 100V/V) of sense voltage for the full-scale current in each application.
- Efficiency and Power Dissipation: At high current levels, the I^2R losses in R_{SENSE} can be significant. Take this into consideration when choosing the resistor value and its power dissipation (wattage) rating. Also, the sense resistor's value might drift if it is allowed to heat up excessively.
- Inductance: Keep inductance low if Isense has a large high-frequency component. Wire-wound resistors have the highest inductance, while metal film is somewhat better. Low-inductance, metal-film resistors are also available. Instead of being spiral- wrapped around a core, as in metal-film or wire-wound resistors, they are a straight band of metal and are available in values under $1Ω$.

Because of the high currents that flow through R_{SENSE}, take care to eliminate parasitic trace resistance from causing errors in the sense voltage. Either use a four-terminal current-sense resistor or use Kelvin (force and sense) PC board layout techniques.

9.2 Dynamic Range Consideration

Although the RSA4081 have fully symmetrical bidirectional V_{SENSE} input capability, the output voltage range is usually higher from REF to V_{cc} and lower from REF to GND (unless the supply voltage is at the lowest end of the operating range). Therefore, the user must consider the dynamic range of current monitored in both directions and choose the supply voltage and the reference voltage (REF) to make sure the output swing above and below REF is adequate to handle the swings without clipping or running out of headroom.

9.3 Power-Supply Bypassing and Grounding

For most applications, bypass V_{CC} to GND with a 0.1μ F ceramic capacitor. In many applications, V_{CC} can be connected to one of the current monitor terminals (RS+ or RS-). Because V_{cc} is independent of the monitored voltage, V_{CC} can be connected to a separate regulated supply.

If V_{CC} will be subject to fast-line transients, a series resistor can be added to the power-supply line of the RSA408X to minimize output disturbance. This resistance and the decoupling capacitor reduce the rise time of the transient. For most applications, 1kΩ in conjunction with a 0.1μF bypass capacitor work well.

The RSA408X require no special considerations with respect to layout or grounding. Consideration should be given to minimizing errors due to the large charge and discharge currents in the system.

9.4 Power Management

The bidirectional capability of the RSA4081 makes it an excellent candidate for use in smart battery packs. In the application diagram (Figure 24), the RSA4081 monitors the charging current into the battery as well as the discharge current out of the battery. The microcontroller stores this information, allowing the system to query the battery's status as needed to make system power-management decisions.

Figure 26. RSA4081 Used in Smart-Battery Application

10 PACKAGE OUTLINE DIMENSIONS

SOP8 (3)

RECOMMENDED LAND PATTERN (Unit: mm)

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.

2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.

3. This drawing is subject to change without notice.

MSOP8 (3)

RECOMMENDED LAND PATTERN (Unit: mm)

NOTE:

1. Plastic or metal protrusions of 0.15mm maximum per side are not included.

2. BSC (Basic Spacing between Centers), "Basic" spacing is nominal.

3. This drawing is subject to change without notice.

11 TAPE AND REEL INFORMATION REEL DIMENSIONS TAPE DIMENSION

NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

NOTE:

1. All dimensions are nominal.

2. Plastic or metal protrusions of 0.15mm maximum per side are not included.

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